# Evaluation of Clipping and Filtering-Based PAPR Reduction in OFDM System

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**Abstract** This paper investigates clipping and filtering techniques in reducing peak average power ratio (PAPR) of Orthogonal Frequency Division Multiplexing (OFDM) system. The concept of OFDM is to split a high speed serial data into parallel data at a lower speed, then the parallel data carried by mutually orthogonal subcarriers. The high of PAPR is one of disadvantages of OFDM system. The high PAPR can damages the form of OFDM and reduces its performance. The purpose of this study is to reduce PAPR using simulation. OFDM was simulated with and without clipping filtering then compared. The methods used to reduce PAPR was clipping and filtering technique. Clipping and filtering technique operates by clipping the output of inverse Fourier transform that exceed the threshold. Graphics PAPR vs. CCDF was used to evaluate the performance of OFDM systems. PAPR for OFDM system using Fourier transform when CCDF =  $10^{-3}$  is 11,2 dB, with classical clipping PAPR was 4,1 dB and PAPR 4,6 dB when with deep clipping.

# Index Terms-discrete wavelet transforms, Fourier transform, PAPR, OFDM

# I. INTRODUCTION

OFDM has several properties which make it an attractive modulation scheme for high speed transmission. One of challenging issues in OFDM is high Peak Average Power Ratio (PAPR). The high PAPR causes the interference and degraded the performance of the system while OFDM signal pass through the amplifier [1,2]. High PAPR force the High Power Amplifier (HPA) to operate in its linear region with wide dynamic range, where the power efficiency is very poor. The poor power efficiency makes the reduction of PAPR more important in OFDM systems. So, clearly it would be desirable to have the average power and peak power values as close together as possible for achieving good performance of OFDM system. Depending upon the reduction in PAPR level and computational complexity, different techniques have been surveyed [3]. There have been many methods in which the PAPR can be reduced in OFDM system, which basically can be divided in three categories, as follows: 1) signal distortion techniques, which reduce the peak amplitudes simply by nonlinearly distorting the OFDM signal at or around the peaks, e.g. clipping and filtering, envelope scaling, peak reduction carrier, peak windowing and peak cancellation; 2) Signal Scrambling Techniques each OFDM symbol with

different scrambling sequences and selecting the sequence that gives the smallest PAPR, e.g. Block Coding Techniques, Block Coding Scheme with Error Correction, Selected Mapping (SLM), Partial Transmit Sequence (PTS), Interleaving Technique, Tone Reservation (TR) and Tone Injection (TI); 3) coding techniques that use a special FEC code set that excludes OFDM symbols with a large PAPR [4,5].

Clipping the signal to a maximum allowed value is a simple technique used to reduce the PAPR of OFDM signals. However, this technique caused degradation of Bit Error Rate (BER) and out-of-band radiation. Clipping does not add extra information to the signal and high peaks occur with low probability so the signal is seldom distorted. At the transmitter, filtering applied to reduce out of-band radiation. A FFT-IFFT pair can be used as filter which is easier to implement than traditional FIR filters and allows the implementation of the clip and filter set several times in order to reduce the peak regrowth that filtering introduces. At the receiver, the signal can be reconstructed for mitigating BER degradation.

The remainder of this paper is organised as follows. Section 2 presents the system model used in this study. Section 3 describes the results and analysis of the performance of OFDM system with and without clipping san filtering technique. Finally, in section 4, a conclusion is drawn.

#### II. SYSTEM MODEL

#### A. DFT-based Implementation

Weinstein and Ebert have shown that OFDM generation and detection can be achieved by employing an IDFT and a DFT at the transmitter and the receiver respectively [6,7]. The structure of a DFT-based base-band OFDM transmitter and receiver is shown in Figure 2. In DFT-based implementation, an OFDM symbol can be generated by taking the IDFT of the complex modulation symbols to be conveyed in each sub-channel, that is given by

$$\mathbf{x}(\mathbf{n}) = \frac{1}{N} \sum_{k=0}^{N-1} \mathbf{X}(k) e^{j2\pi k \mathbf{n}/N},$$
  
k = 0, ...., N - 1,

and the complex modulation symbols conveyed in each subchannel can be detected by taking the DFT of the samples as shown

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j2\pi kn/N},$$
  
n = 0, ...., N - 1.  
(2)

Figure 1 shows the block diagram of the relationship between frequency domain and time domain in an OFDM system [8].



Figure 2 Frequency domain and time domain in an OFDM system.



Figure 1 Structure of a DFT-based baseband OFDM.

#### **B.** Efficient OFDM Implementation

The efficient implementations of an OFDM transmitter and receiver can be constructed in practice by the IFFT and FFT, respectively, that are typically employed in the baseband OFDM basic structure [9], as shown in Figure 3. The input data stream, b bps, is modulated by a type of modulation (M-QPSK, M-QAM), resulting in an M-ary symbol stream, R, where each symbol has in-phase (I) and quadrature (Q) components, each defining one out of M possible constellation points. This symbol stream is passed through a serial-to-parallel converter, which outputs a set of N parallel M-ary symbols,  $S(i)=S(0),S(1),\ldots,S(N-1)$ . These output symbols are transmitted over N sub-channels and correspond to the symbols transmitted over each of the subcarriers. The number of information bits per-OFDM block, denoted as B, depends on the number of sub-carriers, N, and the number of bits per sub-carrier, which depends on the type of modulation used. During each OFDM block, each sub-carrier transmits one symbol. Thus, the N symbols output from the serial-to-parallel converter are the discrete frequency components of the output OFDM modulator, x(n). x(n) is generated by converting its frequency components into time samples by performing an IFFT on the N sub-channels. Each sub-channel modulates a separate carrier through the IFFT modulation block, which actually generates the OFDM block, performing the multicarrier modulation [10].



Figure 3 N-sub-carriers OFDM with IFFT/FFT implementation.

### C. Clipping-Filtering Implementation

The implementation of clipping-filtering for reducing PAPR in OFDM system can be seen in Figure 4.



Figure 4 clipping filtering implementation in OFDM

In this research, the clipping technique was used classical clipping and deep clipping. Both types of clipping have different ways of working in clipping the amplitudes of an OFDM signal.

In classical clipping, it depends on two conditions, which are the amplitude of an OFDM signal will be passed directly when the amplitude of the signal is less than or equal to the threshold value. When signal r is greater than threshold value, signal will experiences clipping amplitude according to level of clipping ratio.

While deep clipping in the process depends on 3 conditions, which are the amplitude of an OFDM signal will be passed directly when the amplitude of the signal is less than or equal to the threshold value. When signal r in  $A_{max} < r \le \frac{1+p}{p} A_{max}$ , then signal clipping process with equation A - p \* (r - A). When  $r > \frac{1+p}{p} A_{max}$ , the signal amplitude will be clipped with the amplitude value "0", where p is the depth level of clipping (deep factor) that has been pre-determined.

## D.Simulation parameter

The parameter used in the simulation can be used in Table 1.

TABLE I SIMULATION PARAMETER

clic	Parameter N frame	Values 5000
ix 🟲	Clipping Ratio (CR)	Classical Clipping = 1,4
		$Deep \ Clipping = 0,6$
	FFT Size	64
	N of data subcarriers	64
	Guard period type	Cyclic prefix
	N bits per OFDM symbol	64
1	N symbol	1
al	N Cyclic prefix	<sup>1</sup> / <sub>4</sub> (N of data subcarriers)
e	Total symbol	N of data subcarriers + N cyclic
n		prefix

# III. RESULTS AND ANALYSIS

The simulation was based on DFT-OFDM model as seen in Figure 2. The performance of OFDM system with and without PAPR reduction scheme was described by CCDF vs. PAPR graphics. The PAPR reduction was simulated using classical and deep clipping-filtering. The three conditions were compared and analyzed.

Simulations were performed using the parameters as seen in table 1. The CCDF graph of the DFT-OFDM simulation can be seen in Figure 4.



Figure 4 Comparison of CCDF using clipping and filtering method

From the result, it can be seen when CCDF  $10^{-3}$ , PAPR of OFDM was 11.2 dB. OFDM system using classical clipping it has a PAPR value of 4.1 dB which capable of reducing PAPR to 7.1 dB. While OFDM system with deep clipping, it has a PAPR value of 4.6 dB which able to reduce the PAPR to 6.6 dB. Classical clipping is able to reduce PAPR more the deep clipping in this simulation.

In accordance with the above simulation results, OFDM with clipping process both classical clipping and deep clipping has a lower PAPR value than OFDM without clipping. That is because in clipping process, the signal amplitude is limited by the threshold and therefore the ratio of peak amplitude to average amplitude will be smaller.

### IV. CONCLUSION

The PAPR problem is one of challenging issues in developing OFDM system. Clipping is a simple solution that results from this paper in order to reduce the effects of high PAPR in OFDM systems. At the transmitter, filtering applied to reduce out of-band radiation. By using clipping and filtering schemes, the performance of OFDM system can be improved.

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